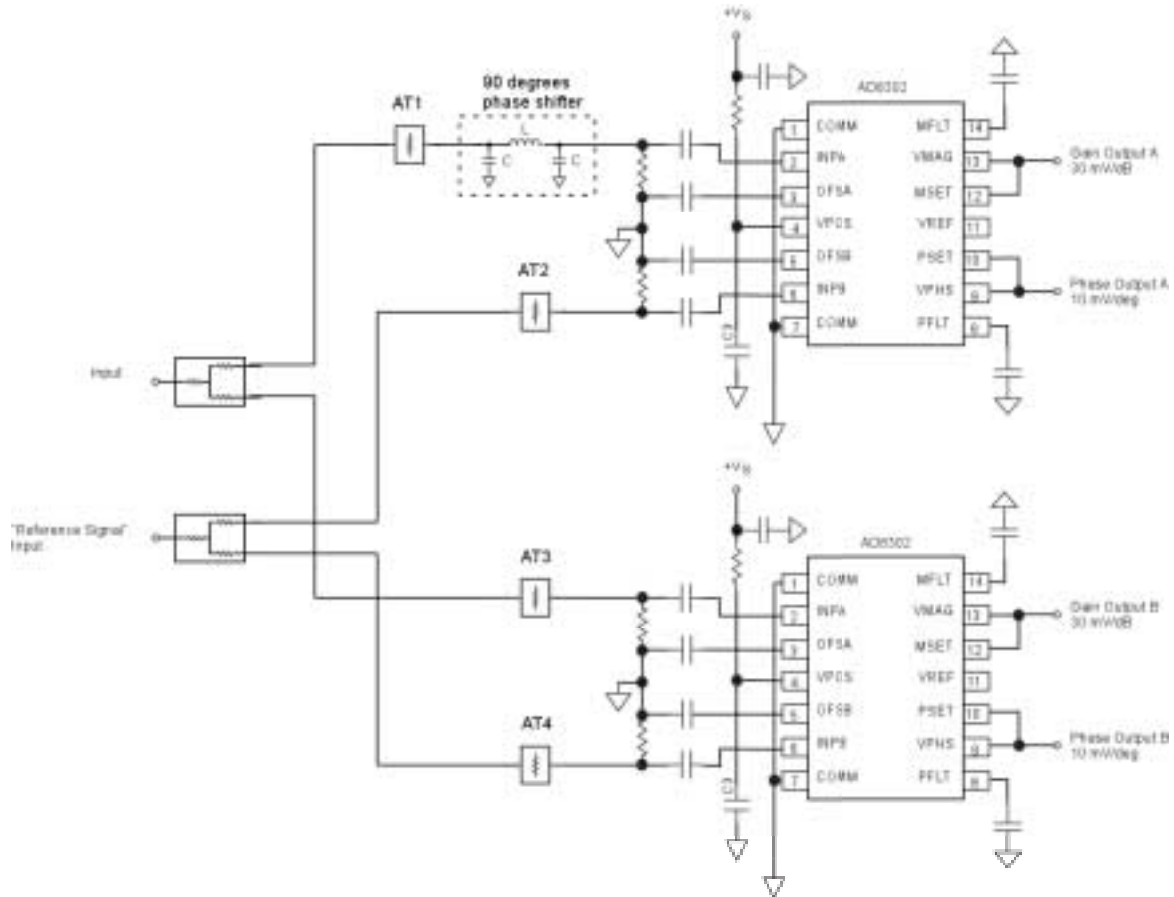
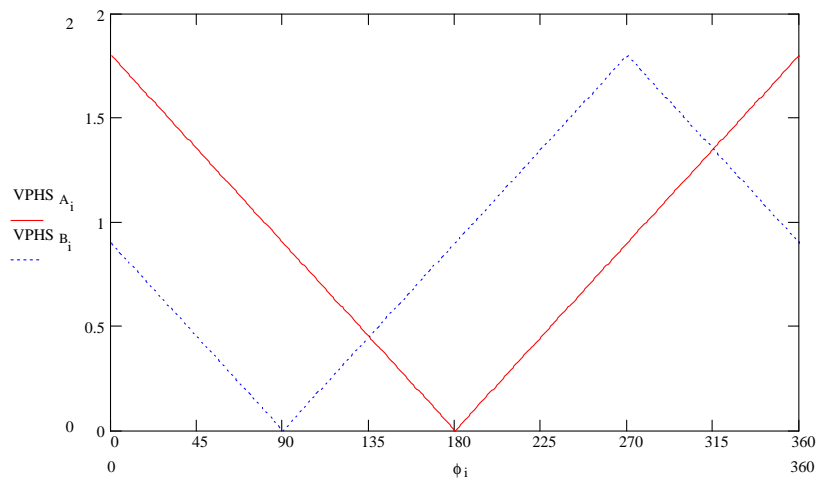


a Full 360° Phase Detection with AD8302 180° Phase Detectors

1. use two AD8302's
 - a. apply identical "reference signal" to each AD8302 INPB
 - b. apply "unknown signal" to INPA of one of the AD8302's
 - c. apply 90°-shifted "unknown signal" to INPA of other AD8302



2. Resulting VPHS outputs are shown below



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3. Determine phase angle, Θ , through following decision tree:

- a. If $V_{PHS_A} \geq (0.75 * V_{REF})$ and $(V_{PHS_B} \leq V_{REF}/2)$ then $\Theta = ((V_{PHS_B} - 0.9)/-10^{-2})$
- b. If $V_{PHS_B} < (0.25 * V_{REF})$ then $\Theta = ((V_{PHS_A} - 0.9)/-10^{-2}) + 90$
- c. If $V_{PHS_A} \leq (0.25 * V_{REF})$ then $\Theta = ((V_{PHS_B} - 0.9)/10^{-2}) + 180$
- d. If $V_{PHS_B} > (0.75 * V_{REF})$ then $\Theta = ((V_{PHS_A} - 0.9)/10^{-2}) + 270$
- e. If $V_{PHS_A} \geq (0.75 * V_{REF})$ and $(V_{PHS_B} \geq V_{REF}/2)$ then $\Theta = ((V_{PHS_B} - 0.9)/-10^{-2}) + 360$

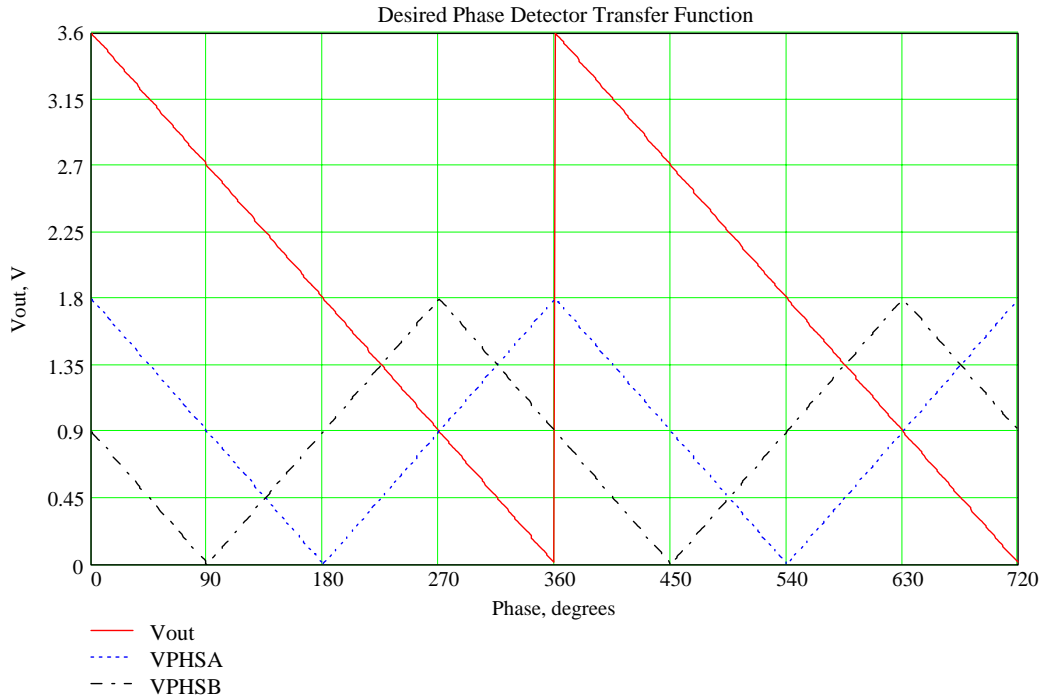
(Note that there are “greater than” ($>$), “greater than or equal to” (\geq), “less than” ($<$) and “less than or equal to” (\leq) inequalities in the decision tree above).

This algorithm could be implemented by digitizing V_{PHS_A} and V_{PHS_B} and using an IF-THEN-ELSE or similar structure in software, then either use this digital value directly or use this digital value to produce an analog output with a DAC.

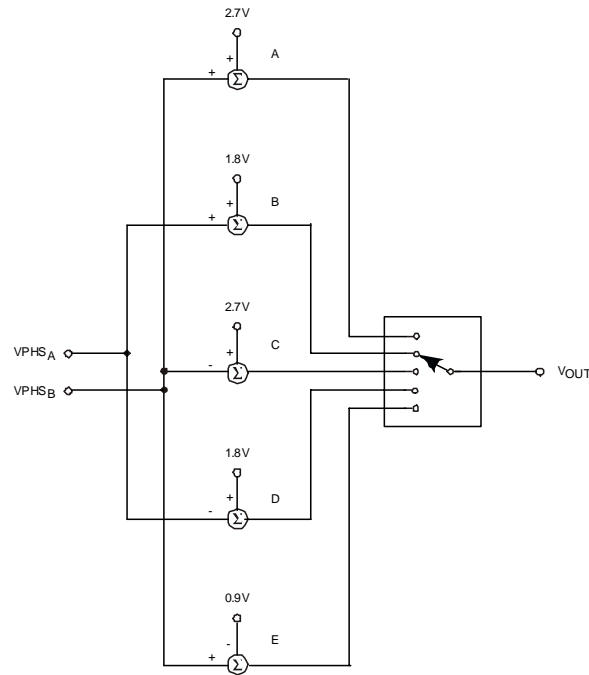
This algorithm uses only the “sweet spot” regions of each V_{PHS} output, i.e., those portions of the curves that are centered around 900 mV.

4. Analog output voltage could be constructed from V_{PHS_A} and V_{PHS_B} according to following decision tree:
 - a. If $\Theta \leq 45^\circ$ then $V_{OUT} = V_{PHS_B} + 2.7 \text{ V}$
 - b. If $(\Theta > 45^\circ)$ and $(\Theta \leq 135^\circ)$ then $V_{OUT} = V_{PHS_A} + 1.8 \text{ V}$
 - c. If $(\Theta > 135^\circ)$ and $(\Theta \leq 225^\circ)$ then $V_{OUT} = 2.7 \text{ V} - V_{PHS_B}$
 - d. If $(\Theta > 225^\circ)$ and $(\Theta \leq 315^\circ)$ then $V_{OUT} = 1.8 \text{ V} - V_{PHS_A}$
 - e. If $\Theta > 315^\circ$ then $V_{OUT} = V_{PHS_B} - 0.9 \text{ V}$

a Full 360° Phase Detection with AD8302 180° Phase Detectors



This waveform could be constructed using the VREF output to produce the constants ($3/2 V_{REF}$, V_{REF} , $1/2 V_{REF}$) that are combined with the appropriate VPHS output using op amp summer and subtractor circuits as shown below



Voltage A is selected when the phase angle, Θ , is $> 0^\circ$ and $\leq 45^\circ$. Voltage B is selected when Θ is $> 45^\circ$ but $\leq 135^\circ$, Voltage C is selected for $135^\circ < \Theta \leq 225^\circ$, Voltage D for

a Full 360° Phase Detection with AD8302 180° Phase Detectors

$225^\circ < \Theta \leq 315^\circ$ and Voltage E for $315^\circ < \Theta \leq 360^\circ$. The selection circuit, which could be a SP5T switch, 1:8 multiplexer, etc., would be controlled by comparators which would determine the proper output voltage selection by comparing $VPHS_A$ and/or $VPHS_B$ to $0.75 * VREF$, $0.25 * VREF$ and each other. This discrete-switching approach might be inferior to a soft-handoff approach that would gradually select transfer the input of the selection circuit.

The summing and subtracting circuits could be constructed from op amp circuits.